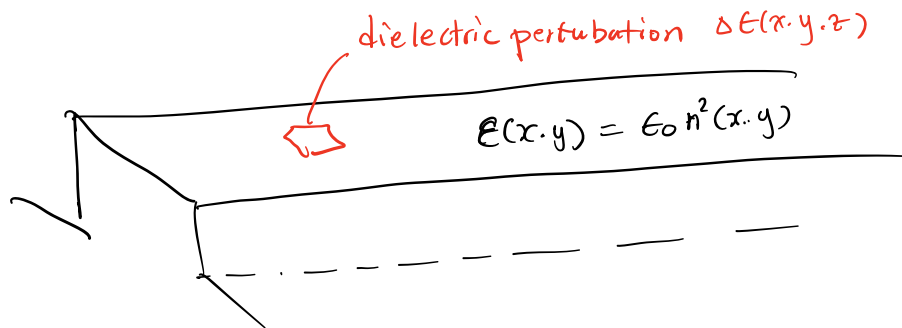


## Lecture 8. Waveguide devices

Learning objectives:

- ① Mode coupling
- ② Waveguide coupling (directional coupler)
- ③ Ring resonator and critical coupling

## 1. Mode coupling (Yariv P603)



when  $\epsilon(x,y)$  is independent of  $(z)$ , modes are independent  
but when dielectric perturbation is applied to the waveguide,  
(bending, surface corrugation), the modes are coupled together.

Eg. incident  $TE_0$  mode, some of its power will be  
transferred to other modes, such as  $TE_1, TE_2, \dots$

Special case,  $\Delta\epsilon(x,y,z) = \Delta\epsilon(x,y)$ .

$$\epsilon(x,y) = \underbrace{\epsilon_a(x,y)} + \Delta\epsilon(x,y)$$

dielectric const. of unperturbed WG.

Let the unperturbed modes be

$$E_m = E_m(x,y) \cdot e^{i(\omega t - \beta_m z)}$$

where  $\epsilon_m(x, y)$  (transverse wave functions) satisfy

$$\nabla_t^2 \epsilon_m(x, y) + [\omega^2 \mu \epsilon_a(x, y) - \beta_m^2] \epsilon_m(x, y) = 0 \quad (1)$$

Apply the perturbation; we have

$$[\nabla_t^2 + \omega^2 \mu \epsilon_a(x, y) + \omega^2 \mu \Delta \epsilon(x, y)] (\epsilon_m + \delta \epsilon_m) = (\beta_m^2 + \delta \beta_m^2) (\epsilon_m + \delta \epsilon_m)$$

neglect the second order term  $\Delta \epsilon \cdot \delta \epsilon_m$  and  $\delta \beta_m^2 \delta \epsilon_m$ ,  
use eq. (1), we have

$$[\nabla_t^2 + \omega^2 \mu \epsilon_a(x, y)] \delta \epsilon_m + \omega^2 \mu \Delta \epsilon \epsilon_m = \beta_m^2 \delta \epsilon_m + \delta \beta_m^2 \epsilon_m \quad (2)$$

To solve (2), we expand  $\delta \epsilon_m(x, y) = \sum_n a_{mn} \epsilon_n(x, y)$

plug in (2), we have

$$\sum_n a_{mn} (\beta_n^2 - \beta_m^2) \epsilon_n(x, y) = (\delta \beta_m^2 - \omega^2 \mu \Delta \epsilon) \epsilon_m(x, y) \quad (3)$$

Scalar-multiply (3) by  $\epsilon_m^*$ , integral over  $x, y$ , use the orthogonal property.

$$\text{i.e.} \quad \int_A \epsilon_m^*(x, y) \epsilon_n(x, y) dx, dy = 0,$$

the left term is gone.

We get.

$$\iint \epsilon_m^* (\delta \beta_m^2 - \omega^2 \mu_0 \Delta \epsilon) \epsilon_m(x, y) = 0$$

$$\Rightarrow \delta \beta_m^2 = \frac{\iint \epsilon_m^* \cdot \omega^2 \mu_0 \Delta \epsilon \epsilon_m dx dy}{\iint \epsilon_m^* \epsilon_m dx dy}$$

Using the orthogonal relation:  $\frac{\beta_m}{2\omega\mu} \iint \epsilon_m^* \epsilon_n^* dx dy = \delta_{mn}$ .

$$\Rightarrow \delta \beta_m = \frac{\omega}{4} \iint \epsilon_m^* \cdot \Delta \epsilon \epsilon_m dx dy.$$

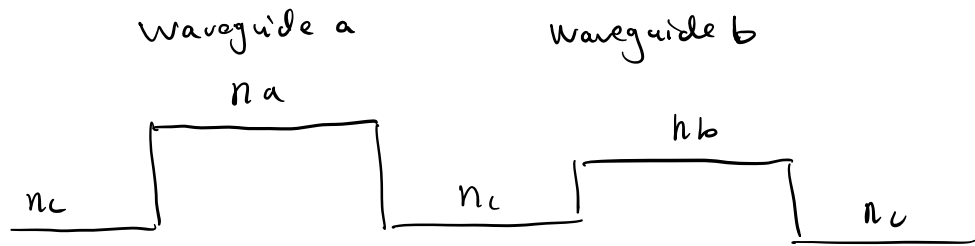
$$a_{mn} = \frac{\omega \beta_n}{2(\beta_m^2 - \beta_n^2)} \iint \epsilon_n^* \cdot \Delta \epsilon(x, y) \epsilon_m dx dy$$

$$a_{mm} = -\frac{\omega}{\delta \beta_m} \iint \epsilon_m^* \Delta \epsilon(x, y) \cdot \epsilon_m dx dy$$

Define  $K_{nm} = \delta \beta_m = \frac{\omega}{4} \iint \epsilon_n^* \Delta \epsilon \epsilon_m dx dy$ .

as the "coupling coefficient" between modes.

## 2 - Waveguide coupling - Directional coupler



Mode:

$$E_a(x, y) e^{i(\omega t - \beta_a z)}$$

$$E_b(x, y) e^{i(\omega t - \beta_b z)}$$

When they are closer (not too close!),  
the general wave propagation in the coupled waveguide is

$$\vec{E}(x, y, z, t) = A(z) E_a(x, y) e^{i(\omega t - \beta_a z)} + B(z) E_b(x, y) e^{i(\omega t - \beta_b z)}$$

Note:

- ① if distance between a. b is infinite,  
 $A(z), B(z)$  do not depend on  $z$ ,

Index distribution  $n^2(x, y)$

$$n^2(x, y) = \begin{cases} n_a^2 & \text{core a} \\ n_b^2 & \text{core b} \\ n_c^2 & \text{elsewhere} \end{cases}$$

$$\text{define } \Delta n_a^2(x, y) = \begin{cases} n_a^2 - n_c^2 & \text{core a} \\ 0 & \text{elsewhere} \end{cases}$$

$$\Delta n_b^2(x, y) = \begin{cases} n_b^2 - n_c^2 & \text{core b} \\ 0 & \text{elsewhere} \end{cases}$$

$$n_s^2(x, y) = n_c^2$$

The index profile:  $n^2(x, y) = n_s^2(x, y) + \Delta n_a^2(x, y) + \Delta n_b^2(x, y)$

So, the wave equation for the composite waveguide:

$$\left( \frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \frac{\partial^2}{\partial z^2} + \frac{\omega^2}{c^2} [n_s^2(x, y) + \Delta n_a^2(x, y) + \Delta n_b^2(x, y)] \right) \vec{E} = 0 \leftarrow$$

$$\text{where } \vec{E} = A(z) E_a(x, y) e^{i(\omega t - \beta_a z)} + B(z) E_b(x, y) e^{i(\omega t - \beta_b z)}$$

Goal: solve  $A(z)$ ,  $B(z)$ !

Assuming slow variation of mode amplitudes over  $z$ ,  
we get

$$\begin{aligned}
 & -2i\beta_a \frac{dA}{dz} \epsilon_a e^{i(\omega t - \beta_a z)} - 2i\beta_b \frac{dB}{dz} \epsilon_b e^{i(\omega t - \beta_b z)} \\
 & = -\frac{\omega^2}{c^2} \Delta n_b^2(x,y) A \epsilon_a e^{i(\omega t - \beta_a z)} - \frac{\omega^2}{c^2} \Delta n_a^2(x,y) B \epsilon_b e^{i(\omega t - \beta_b z)}
 \end{aligned}$$

Take scalar product with  $\epsilon_a^*(x,y)$   $\epsilon_b^*(x,y)$  and integrate over  $x,y$  plane, we get

$$\begin{aligned}
 \frac{dA}{dz} &= -ik_{ab} B e^{i(\beta_a - \beta_b)z} - ik_{aa} A \\
 \frac{dB}{dz} &= -ik_{ba} A e^{-i(\beta_a - \beta_b)z} - ik_{bb} B
 \end{aligned}$$

} ④  
coupled mode eq. (CME)

where  $k_{ab} = \frac{\omega}{4} \epsilon_0 \iint \epsilon_a^* \Delta n_a^2(x,y) \epsilon_b dx dy$

$$k_{ba} = \frac{\omega}{4} \epsilon_0 \iint \epsilon_b^* \Delta n_b^2(x,y) \epsilon_a dx dy$$

$$k_{aa} = \frac{\omega}{4} \epsilon_0 \iint \epsilon_a^* \Delta n_b^2(x,y) \epsilon_a dx dy$$

$$k_{bb} = \frac{\omega}{4} \epsilon_0 \iint \epsilon_b^* \Delta n_a^2(x,y) \epsilon_b dx dy$$

Comments:

- ①  $k_{ab}, k_{ba}$  are exchange coupling between 2 waveguides
- ②  $k_{aa}, k_{bb}$  result from perturbation to one waveguide due to the presence of another

③  $K_{ab} = K_{ba}^*$ , ensure the conservation of energy.

If assuming

$$E(x, y, z, t) = A(z) \epsilon_a e^{i(\omega t - (\beta_a + K_{aa})z)} + B(z) \epsilon_b e^{i(\omega t - (\beta_b + K_{bb})z)}$$

CME ④ reduces to

$$\begin{cases} \frac{dA}{dz} = -i K_{ab} B e^{i2\delta z} \\ \frac{dB}{dz} = -i K_{ba} A e^{-i2\delta z} \end{cases}, \quad 2\delta = (\beta_a + K_{aa}) - (\beta_b + K_{bb}) \quad \text{⑤}$$

Assuming  $K_{ab} = K_{ba} = k$ ,  $s = \sqrt{k^2 + \delta^2}$ , solution of ⑤ is

$$\begin{cases} A(z) = A_0 e^{i\delta z} \left( \cos sz - i \frac{\delta}{s} \sin sz \right) \\ B(z) = -i A_0 e^{-i\delta z} \frac{k}{s} \sin sz \end{cases}$$

In terms of power,  $P_a(z) = |A(z)|^2$ ,  $P_b(z) = |B(z)|^2$

$$\begin{cases} P_a(z) = P_0 - P_b(z) \end{cases} \quad \text{①}$$

$$\begin{cases} P_b(z) = P_0 \frac{k^2}{k^2 + \delta^2} \sin^2 \sqrt{k^2 + \delta^2} z \end{cases} \quad \text{②}$$

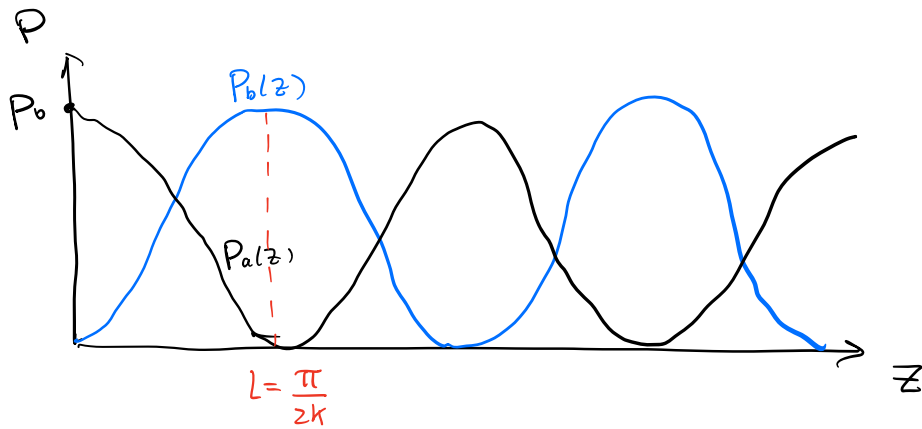
where  $P_0 = |A(0)|^2 = A_0^2$  is the input power at  $z=0$



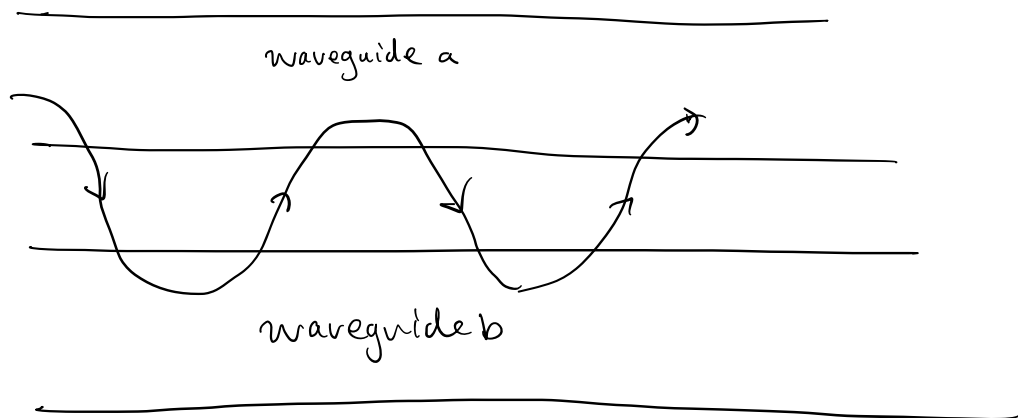
## Comments

1) When  $\delta = 0$ , ( $\beta_a = \beta_b$ ), (2) becomes

$$\begin{cases} P_a(z) = P_0 - P_b(z) \\ P_b(z) = P_0 \sin^2 kz \end{cases}$$



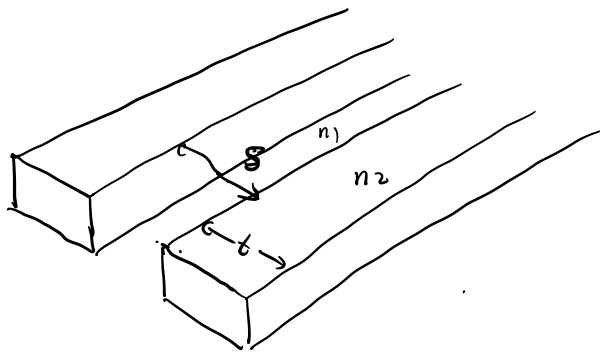
Directional coupler:



by controlling the coupling length  $L$ ,  
we can control  $P_a(z)/P_b(z)$

2) For  $\delta \neq 0$ , the maximum fraction of power that can be transferred is  $\frac{\kappa^2}{\kappa^2 + \delta^2}$

For two identical channel waveguides



If modes are well confined ( $t \gg 2q_1$ )

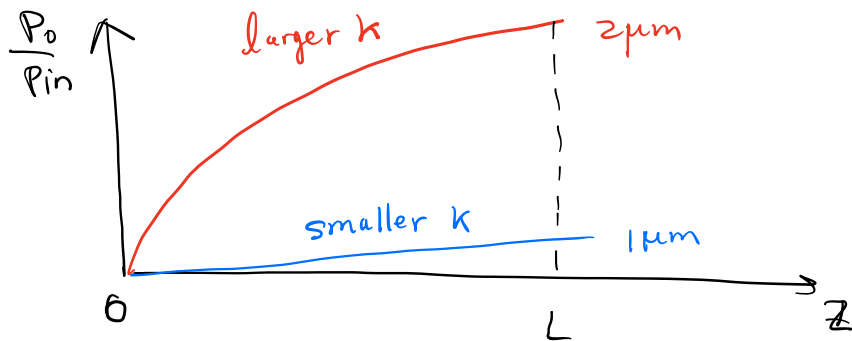
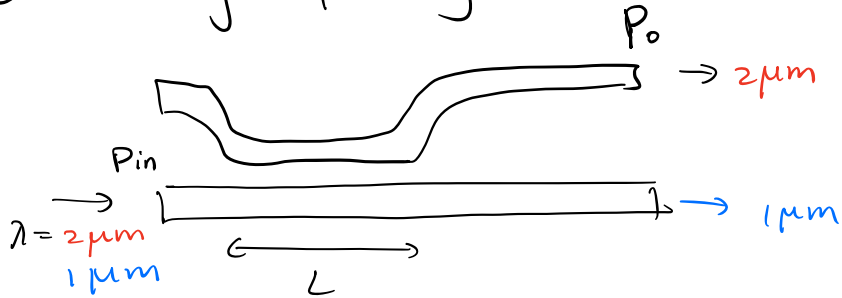
$$\kappa = \frac{2h^2 q_1 e^{-\gamma s}}{\beta t (h^2 + q_1^2)} \left( \frac{2\pi}{\lambda} \right)^2 (n_2^2 - n_1^2)$$

At  $\lambda = 1 \mu\text{m}$ ,  $t, s \sim 3 \mu\text{m}$ ,  $\Delta n = 5 \times 10^{-3}$ ,

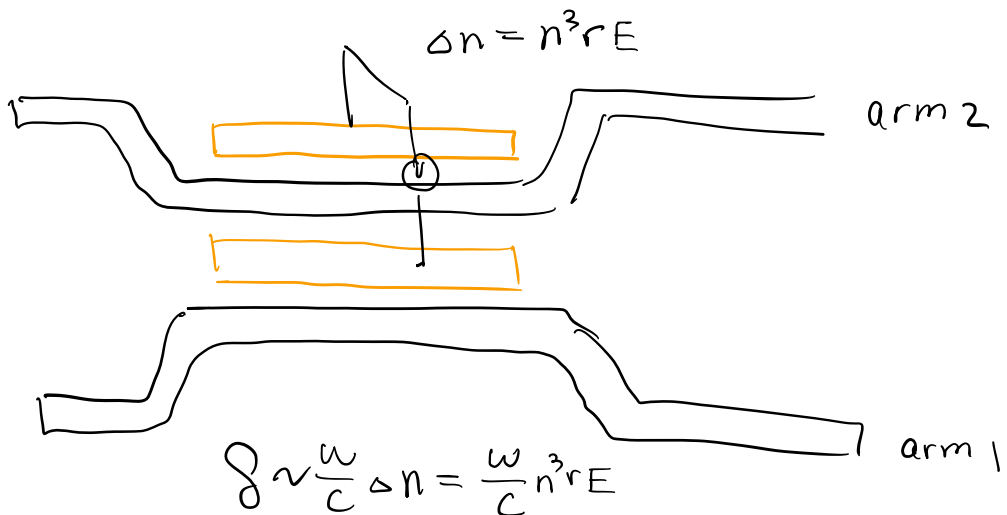
$\kappa = 5 \text{ cm}^{-1}$ , coupling length  $L \sim \kappa^{-1} = 2 \text{ mm}$ .

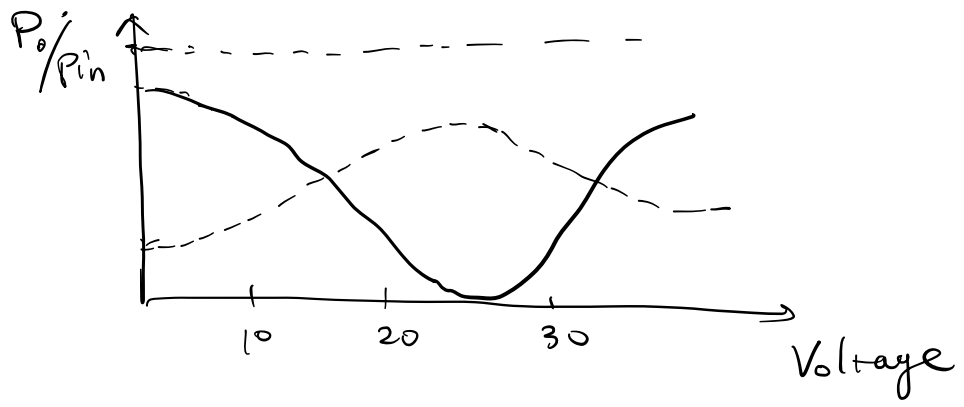
# Applications

## ① Wavelength filtering



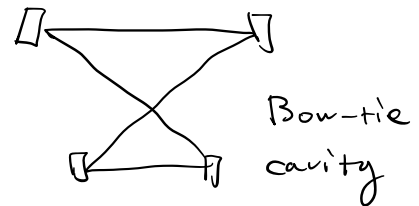
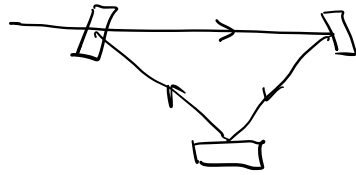
## ② Electro-optic switch



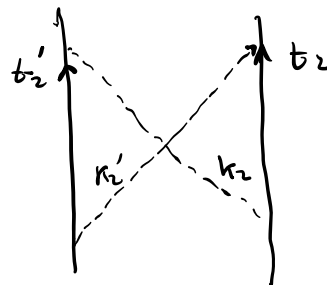
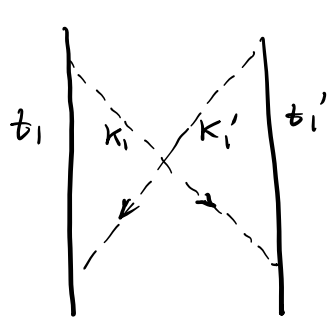
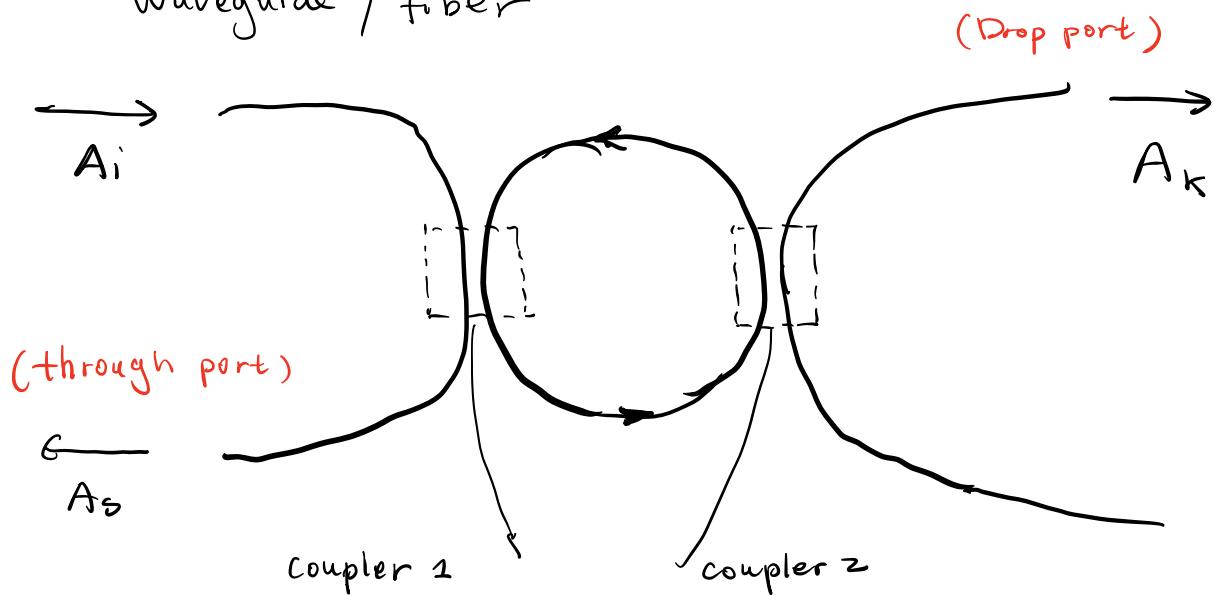


### 3. Ring resonators and critical coupling (Yariv, P185)

Free space ring resonators :



Waveguide / fiber



Define ① straight-through coupling coefficient  $t_1, t_1', t_2, t_2'$

② Cross-coupling coefficient  $k_1, k_1', k_2, k_2'$

Model the ring as Fabry-Perot Etalon, Then,

$$A_s = t_1 A_i + k_1 k_1' t_2' e^{-i\delta} A_i + k_1 k_1' t_2' t_1' t_2' e^{-2i\delta} + \dots$$

$$= \left\{ t_1 + k_1 k_1' t_2' e^{-i\delta} \left[ 1 + t_1' t_2' e^{-i\delta} + (t_1' t_2' e^{-i\delta})^2 + \dots \right] \right\} A_i$$

$$A_k = k_1 k_2' e^{-i\delta/2} \left[ 1 + t_1' t_2' e^{-i\delta} + (t_1' t_2' e^{-i\delta})^2 + \dots \right] A_i$$

where  $\delta = \frac{2\pi n_{\text{eff}} d}{\lambda}$  ← circumference of the ring.

Summing up the geometric series,

$$\sigma = \frac{A_s}{A_i} = \frac{t_1 + (k_1 k_1' - t_1 t_1') t_2' e^{-i\delta}}{1 - t_1' t_2' e^{-i\delta}}$$

$$\gamma = \frac{A_k}{A_i} = \frac{k_1 k_2' e^{-i\delta/2}}{1 - t_1' t_2' e^{-i\delta}}$$

When there is loss due to scattering, absorption and bending.

$$\sigma = \frac{A_s}{A_i} = \frac{t_1 + (k_1 k_1' - t_1 t_1') t_2' e^{-i\delta} e^{-2d}}{1 - t_1' t_2' e^{-i\delta} e^{-2d}} \quad (\text{T from input to through})$$

$$\chi = \frac{A_r}{A_i} = \frac{k_1 k_2 e^{-i\frac{\delta}{2}} e^{-2d}}{1 - t_1' t_2' e^{-i\delta} e^{-2d}} \quad (\text{T from input to drop})$$

which is quite similar to F-P etalon (cavity).

Q: How to determine  $(t_1, t_1', t_2, t_2')$  and  $(k_1, k_1', k_2, k_2')$ ?

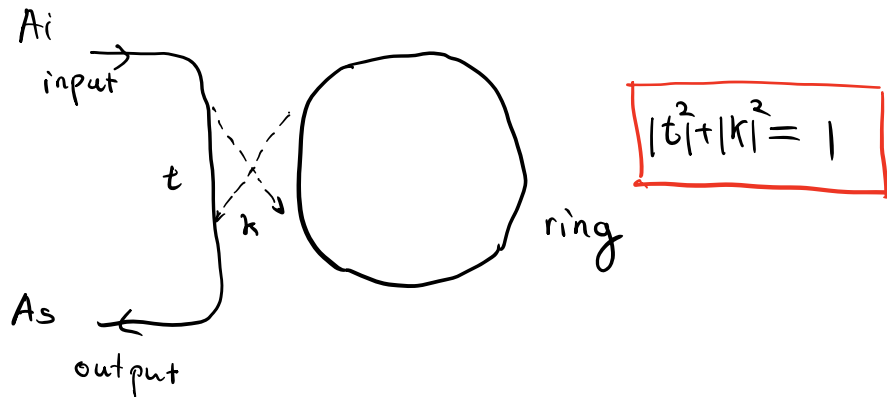
A: They are not independent of each other. They are related by reciprocity, conservation of energy and time-reversal symmetry.

$$\text{i.e. } \begin{cases} t_1^2 + k_1^2 = t_1'^2 + k_1'^2 = 1 \\ t_1 t_1' - k_1 k_1' = -1 \end{cases} \quad (\text{Yariv P186~188})$$

Then we have:

$$\sigma = \frac{t_1 + t_2' e^{-i\delta} e^{-2d}}{1 - t_1' t_2' e^{-i\delta} e^{-2d}}$$

Assuming  $k_2 = 0$ ,  $t_2 = -1$ , we have



$$\sigma = \frac{A_s}{A_i} = \frac{t - ae^{-i\delta}}{1 + t'ae^{-i\delta}} = \frac{t - ae^{-i\delta}}{1 - t^*ae^{-i\delta}}$$

where  $t' = -t_1^* = -t^*$ ,  $a = \exp(-2\alpha)$

$$|\sigma|^2 = \left| \frac{A_s}{A_i} \right|^2 = \frac{a^2 + t^2 - 2at\cos\delta}{1 + a^2t^2 - 2at\cos\delta}$$

When at resonance ( $\delta = 2m\pi$ ,  $m = 1, 2, \dots$ )

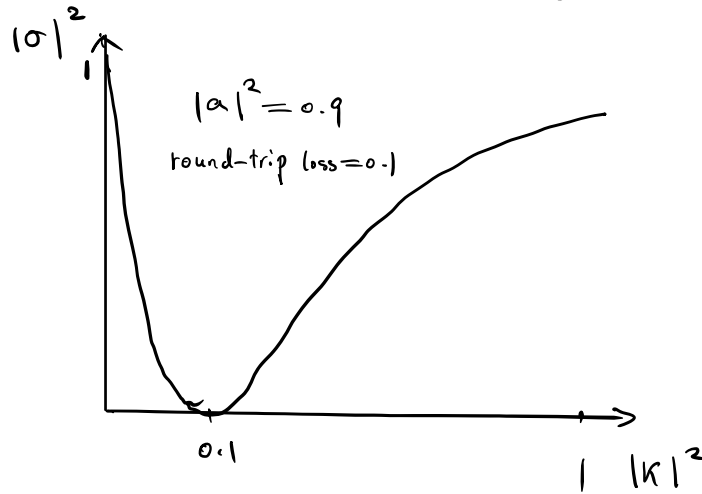
$$|\sigma|^2 = \left| \frac{A_s}{A_i} \right|^2 = \frac{(a-t)^2}{(1-at)^2}$$



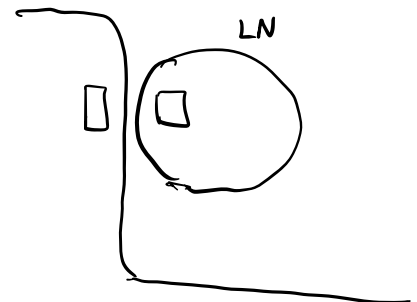
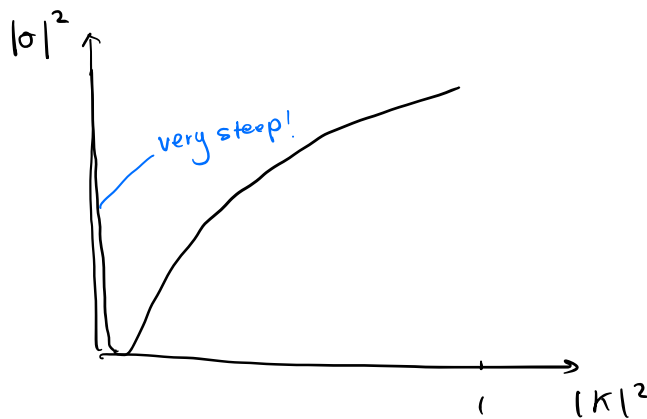
Comments:

①  $|S|>0$  when  $a = t = \sqrt{1-|k|^2}$ , i.e.  $|k|^2 = 1 - |a|^2 = \text{round-trip loss}$

This is known as critical coupling condition



② For high Finesse, high Q-resonators,  $|a| \sim 1$ .  
round-trip loss =  $1 - |a| \approx 0$ .



A small modulation of  $|k|$  will modulate the transmission from zero to unity.  $\rightarrow$  Applications, high modulation depth modulators.